

- 11) Veröffentlichungsnummer:
- (11) Publication number:
- Numéro de publication:

0 928 427

Internationale Anmeldung veröffentlicht durch die Weltorganisation für geistiges Eigentum unter der Nummer:

WO 98/13704 (art.158 des EPÜ).

International application published by the World
Intellectual Property Organisation under number:

 $WO\ 98/13704$ (art.158 of the EPC).

Demande internationale publiée par l'Organisation Mondiale de la Propriété sous le numéro:

WO 98/13704 (art.158 de la CBE).

ς .

PCT

WORLD INTELLECTUAL PROPERTY ORGANIZATION International Bureau



INTERNATIONAL APPLICATION PUBLISHED UNDER THE PATENT COOPERATION TREATY (PCT)

(51) International Patent Classification 6:

G01S 7/282

(11) International Publication Number:

WO 98/13704

A1

(43) International Publication Date:

2 April 1998 (02.04.98)

(21) International Application Number:

PCT/GB97/02513

(22) International Filing Date:

18 September 1997 (18.09.97)

(30) Priority Data:

9620192.6

27 September 1996 (27.09.96)

(71) Applicant (for all designated States except US): GEC- MAR-CONI LIMITED [GB/GB]; The Grove, Warren Lane, Stanmore, Middlesex HA7 4LY (GB).

(72) Inventors; and

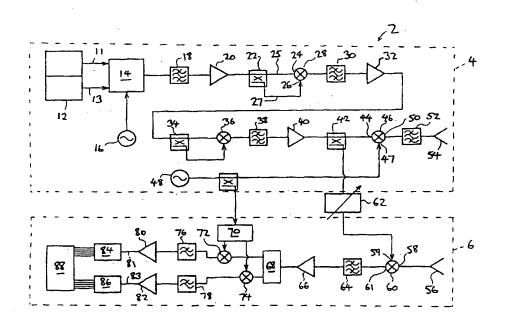
- (75) Inventors/Applicants (for US only): TAIT, Peter, Donald, Fraser [GB/GB]; 176 Rainsford Road, Chelmsford, Essex CM1 2PD (GB). KYTE, Adrian, Peter [GB/GB]; 28 Hill View Road, Springfield, Chelmsford, Essex CM1 7RX (GB). STEWARD, Peter, James [GB/GB]; 8 Hunters Way, Springfield, Chelmsford, Essex CMI 6FL (GB). SHEP-HARD, David, John [GB/GB]; 58 Longstomps Avenue, Chelmsford, Essex CM2 9LA (GB). FFRENCH, Timothy, Edward [GB/GB]; 28 Hill View Road, Springfield, Chelmsford, Essex CM1 7RX (GB).
- (74) Agent: WALKER, Andrew, John; GEC Patent Dept., Waterhouse Lane, Chelmsford, Essex CM1 2QX (GB).

(81) Designated States: AL, AM, AT, AU, AZ, BA, BB, BG, BR, BY, CA, CH, CN, CU, CZ, DE, DK, EE, ES, FI, GB, GE, GH, HU, IL, IS, JP, KE, KG, KP, KR, KZ, LC, LK, LR, LS, LT, LU, LV, MD, MG, MK, MN, MW, MX, NO, NZ, PL, PT, RO, RU, SD, SE, SG, SI, SK, SL, TJ, TM, TR, TT, UA, UG, US, UZ, VN, YU, ZW, ARIPO patent (GH, KE, LS, MW, SD, SZ, UG, ZW), Eurasian patent (AM, AZ, BY, KG, KZ, MD, RU, TJ, TM), European patent (AT, BE, CH, DE, DK, ES, FI, FR, GB, GR, IE, IT, LU, MC, NL, PT, SE), OAPI patent (BF, BJ, CF, CG, CI, CM, GA, GN, ML, MR, NE, SN, TD, TG).

Published

With international search report.

(54) Title: RADAR SYSTEMS



(57) Abstract

A coherent radar detection system (2) comprises a radar signal transmitter (4) and a correlation receiver (6). The transmitter (4) comprises a waveform generator which generates a signal at an intermediate frequency. The signal is divided into two divided signals by a coupler, and then the divided signals are mixed together in a mixer to generate an output signal which has a wider bandwidth than the intermediate frequency. The bandwidth can be increased further by repeating the coupler/mixer stage. The system generates very wide bandwidth signals coherently, allowing coherent processing in high resolution range gates.

FOR THE PURPOSES OF INFORMATION ONLY

Codes used to identify States party to the PCT on the front pages of pamphlets publishing international applications under the PCT.

AL	A Ibania	ES	Spain	LS	Lesotho	SI	Slovenia
AM	Armenia	FI	Finland	LT	Lithuania	SK	Slovakia
AT	Austria	FR	France	LU	Luxembourg	SN	Senegal
AU	Australia	GA	Gabon	LV	Latvia	SZ	Swaziland
AZ	Azerbaijan	GB.	United Kingdom	MC	Monaco	TD	Chad
BA	Bosnia and Herzegovina	GE	Georgia	MD	Republic of Moldova	TG	Togo
ВВ	Barbados	GH	Ghana	MG	Madagascar	TJ	Tajikistan
BE	Belgium	GN	Guinea	MK	The former Yugoslav	TM	Turkmenistan
BF	Burkina Faso	GR	Greece		Republic of Macedonia	TR	Turkey
BG	Bulgaria	HU	Hungary	ML	Mali	TT	Trinidad and Tobago
BJ	Benin	IE	Ireland	MN	Mongolia	UA	Ukraine
BR	Brazil	II.	Israel	MR	Mauritania	UG	Uganda
BY	Belarus	18	Iceland	MW	Malawi	US	United States of America
CA	Canada	IT	. Italy	MX	Mexico	UZ	Uzbekistan
CF	Central African Republic	JP	Japan	NE	Niger	VN	Viet Nam
CG	Congo	KE	Kenya	NL	Netherlands	YU	Yugoslavia
СН	Switzerland	KG	Kyrgyzstan	NO	Norway	ZW	Zimbabwe
CI	Côte d'Ivoire	KP	Democratic People's	NZ	New Zealand		
CM	Cameroon		Republic of Korea	PL	Poland		•
CN	China	KR	Republic of Korea	PT	Portugal		
CU	Cuba	ΚZ	Kazakstan	RO	Romania		
CZ	Czech Republic	LC	Saint Lucia	ŔŪ	Russian Federation		
DE	Germany	LI	Liechtenstein	SD	Sudan		
DK	Denmark	LK	Sri Lanka	SE	Sweden		
EE	Estonia	LR	Liberia	SG	Singapore		

10

15

20

RADAR SYSTEMS

This invention relates to radar systems and is particularly, but not exclusively, related to coherent radar systems.

The range resolution achievable using radar techniques is determined by the signal bandwidth which can be generated and detected. The wider the bandwidth the better (smaller) the resolution. Current techniques provide a maximum of about 1GHz bandwidth which corresponds to about 15cm range resolution. Although systems using impulse radar techniques can provide more bandwidth and thus higher range resolution, they cannot be used as coherent radar systems and thus cannot discriminate between moving and stationary targets. Furthermore in impulse radar it is difficult to control the spectrum shape, which causes range sidelobes resulting in false target detections. Also, because impulse radar operates at low frequencies, it is impossible to transmit a narrow beam which means that such systems are unsuitable for long range radar in which angular discrimination is required.

In some applications high range resolution detection and motion measurement is required. This would be necessary if a radar system was to detect a target which is moving relatively with respect to a background since the detectable amplitude of a return signal from the target would be swamped by the return signal from the background. Examples of such applications include detecting an aircraft against a background of terrain and detecting a particular moving object among a number of moving objects.

15

20

It is an object of the invention to provide a radar system having an improved radar resolution.

According to a first aspect the invention provides radar apparatus comprising waveform generator means for generating a signal at an intermediate frequency, a coupler for dividing the signal into two divided signals, a mixer for mixing the two divided signals together to generate an output signal and transmission means for transmitting the output signal as a radar signal in which the output signal has a wider bandwidth than the signal at the intermediate frequency.

Preferably the waveform generator means generates a wideband signal at the intermediate frequency. Preferably the intermediate frequency is at 1GHz or higher. In one embodiment the intermediate frequency is less than 1 GHz, for example 500MHz.

Preferably the waveform generator means comprises a waveform generator and a vector modulator. The vector modulator may be driven by a local oscillator at a suitable intermediate frequency.

Preferably the waveform generator comprises a two channel digital waveform generator. Preferably the two waveform channels are fed to in-phase and quadrature inputs of the vector modulator to produce a waveform at the intermediate frequency with twice the bandwidth of each waveform channel.

Preferably the output of the vector modulator is filtered to remove spurious signals and amplified by an amplifier before entering the coupler. Preferably, the gain of the amplifier and the coupling value of the coupler may be chosen to give suitable input levels to the mixer.

Preferably a path between a first output of the coupler and a first input of the mixer and a path between a second output of the coupler and a second input of the mixer are of equal length, or of equal phase length.

Preferably the signals which are fed into the mixer have power levels which are substantially different. One may be 15 to 20 dB lower in power level than the other.

Preferably the output signal from the mixer is filtered to remove spurious signals.

Preferably the frequency of the output signal is twice the intermediate frequency of the signal.

Preferably the bandwidth of the output signal is twice the bandwidth of the intermediate frequency signal.

Use of the word "frequency" in relation to a signal or waveform may be reference to the centre frequency of the signal or waveform.

The output signal may be transmitted from an antenna as a radar signal or may be sent to one or more additional coupler and mixer combinations to increase the bandwidth further. The or each additional mixer may produce a new output signal. Suitable amplification and filtering may be provided in relation to each coupler and mixer combination.

Preferably there are two mixers. In an embodiment in which there are two mixers which each double the frequency and bandwidth, the bandwidth of the intermediate signal can be increased by a factor of four. More than two mixers may be present.

15

20

Preferably the apparatus includes a sufficient number of mixers to provide a range resolution which is less than 10cm. It may be in the order of a few cm.

Preferably the waveform is a frequency modulated waveform, for example frequency modulated continuous wave (FMCW) or a within pulse chirp. The output signal may also have FMCW modulation or a within pulse chirp.

The output signal which is ultimately produced may be transmitted from an antenna or may first be translated to a different centre frequency using a conventional mixer.

10

15

20

5

Preferably the apparatus is incorporated into a radar detection system comprising the radar apparatus and means for receiving and processing radar signals.

Preferably the means for receiving radar signals is a correlation receiver, in which the received signal (reflected from a target) is correlated with a reference signal. Conveniently the reference signal is a delayed sample of the output signal which was transmitted.

Correlation may be performed by mixing the received signal and the reference signal in a wideband mixer and integrating the mixer output for a time corresponding to the length of the transmitted waveform. Conveniently this integration may be performed by a low pass filter with a suitable cut-off frequency.

Preferably the correlation receiver retains the phase information of the received signal by performing the correlation in In-phase (I) and Quadrature (Q) channels. Alternatively an offset

frequency may be employed so that the correlation mixer output is at an intermediate frequency, and the subsequent down-conversion to baseband may be performed in I and Q channels.

Preferably the radar system uses coherent processing techniques. It may use Doppler techniques to measure the speed or velocity of a target. It may measure the speed or velocity of the target relative to that of a background.

According to a second aspect the invention provides a method of generating a radar signal comprising the steps of:

generating a signal having an intermediate frequency;

dividing the signal into two divided signals;

mixing the divided signals to generate an output signal having a wider bandwidth than the intermediate frequency signal; and

transmitting the output signal as a radar signal.

15

20

5

Preferably mixing of the divided signals is performed by a mixer. The output signal produced by the mixer may be transmitted from an antenna as a radar signal or may undergo at least one additional mixing step. The or each additional mixing step may produce a new output signal. Preferably there are two mixing steps. There may be more than two mixing steps. The or each mixing step may double the bandwidth of the waveform. The or each mixing step may double the frequency of the waveform.

Preferably the method is used in a radar detection method comprising the steps of: generating the radar signal, receiving a reflected radar signal and processing the reflected radar signal.

WO 98/13704 PCT/GB97/02513

6

Preferably a received signal may be correlated with a reference signal. The received signal may be match filtered against the reference signal. Conveniently the reference signal is identical to the output signal which was transmitted. Received signals may be match filtered against the reference signal in in-phase and quadrature channels.

5

The invention may be used in a pulse radar system, although it could be used on a CW signal which has suitable wideband modulation.

10

The invention may be applied to short range and long range radar detection. It may be used in ground penetration radar or in a radar for the detection, imaging or tracking of moving objects.

An embodiment of the invention will now be described, by way of example only, with reference to the accompanying Figure. This shows a radar system 2 comprising a transmitter 4 and a receiver 6.

15

20

In the transmitter 4, a digital waveform generator 12 produces cosine 11 and sine versions 13 of a chirp pulse waveform. Each of the cosine and sine versions are of 500MHz bandwidth. They are fed to the in-phase (I) and quadrature (Q) ports of a vector modulator 14 driven from a 4GHz local oscillator 16. The vector modulator produces a waveform output which is a chirp pulse with a 1GHz bandwidth centred at 4GHz, that is, the frequency within the pulse varies from 3.5GHz to 4.5GHz. The waveform output is passed through a bandpass filter 18 to remove any undesired out-of-band spurious signals. As a result, a waveform having an intermediate frequency is produced which can be processed further.

10

15

20

The waveform is amplified by an amplifier 20. A coupler 22 divides the waveform into two signals which are used to drive an intermediate frequency (I.F.) port 24 and a local oscillator (L.O.) port 26 of a wideband mixer 28. The coupler has a coupling value which is chosen so that the signal at the L.O. port 26 is a high level signal with power in the range specified for that port, and the signal at the I.F. port 24 is low level signal, for example 15 to 20dB lower in power than the signal used to drive the L.O. port 26. The two signals take respective paths 25, 27 from the coupler to the ports of the mixer 28. The time delay and phase length of the two paths 25, 27 should be matched so that an output signal is produced which has the same time duration as the two input signals and the centre frequency and within-pulse modulation bandwidth are twice the input frequency and bandwidth. The output signal now has a within pulse chirp of 7 to 9GHz. It may be necessary to include an adjustable line length in one or both of the paths 25, 27 from the coupler 22 to the mixer 28.

The coupler 22 and mixer 28 in combination act as a multiplier which produces an output signal having (in this embodiment) a bandwidth and frequency twice that of the waveform which was originally available at the output of the vector modulator 14.

The output signal passes through a filter 30 and is filtered to remove any undesired out-of-band spurious signals generated by the mixer 28. It is then amplified by an amplifier 32 to a level suitable for a further multiplication step. This is performed by a second combination of a coupler 34 and mixer 36 which operates similarly to the combination described above. Again the frequency and bandwidth are both multiplied by a factor of two.

The second combination of coupler and mixer (serving as a second bandwidth multiplier)

produces an output signal which has a within pulse chirp of 14 to 18GHz. The output signal from the second mixer 36 is filtered and amplified by a filter 38 and an amplifier 40. The output signal is passed through a coupler 42 and a sample of the output signal is coupled off to be used in the receiver 6.. This will be discussed below. A portion of the output signal which remains after coupler 42 is fed to a LO port 44 of a wideband mixer 46. An IF port 47 of the mixer 46 is driven from an 8GHz CW oscillator 48, to produce an output signal at a RF port 50 with a within pulse chirp modulation from 22 to 26GHz. This output signal is passed through a bandpass filter 52 to remove spurious signals and transmitted as a radar signal from a transmitting horn antenna 54.

10

15

5

Received signals which have been reflected by a target are received by the receiver 6. This has a receiving horn antenna 56 which feeds received signals to an RF port 58 of a wideband mixer 60. The sample of the output signal taken from coupler 42 is passed through a variable delay 62 which serves as a range gate control and is then fed to a LO port 59 of the wideband mixer 60. The output from an IF port 61 of the mixer 60 is then filtered by a bandpass filter 64 centred at 8GHz, before being passed through a low noise amplifier 66. The amplifier 66 together with the losses of the components which precede it determines the noise figure of the receiver. If a more sensitive receiver 6 is required a low noise amplifier should be included between the receiving horn antenna 56 and the wideband mixer 60.

20

Down-conversion to baseband is performed in a quadrature mixer configuration comprising a zero degree splitter 68, a 90° coupler 70 and two mixers 72, 74. The mixers are matched in gain and phase to produce I and Q signals at baseband. The baseband signals are integrated in low pass filters 76, 78 which are matched to the chirp pulse length and amplified by amplifiers 80,

10

15

20

82 to a level suitable for the analogue-to-digital convertors 84, 86 (ADC) which digitise the received signals. Digitised signals are stored in computer memory 88 for further off-line signal processing which may include Doppler processing.

The receiver acts as a correlator with a multiplication process performed in the wideband mixer 60 at the RF stage and an integration process performed in the low pass filters 76, 78 at baseband. The power level at the output of the correlation receiver, represented by the sum of the squares of the instantaneous amplitudes of outputs 81 and 83 is at a maximum value if the variable delay which is added to the sample of the output signal corresponds to the round trip range delay to the target. In this case the receiver outputs 81, 83 have a frequency corresponding to the Doppler frequency of the target. At all other target ranges, that is those for which the variable delay does not correspond to the round trip range delay, the output signal level is low, given by the off peak level of the waveform auto-correlation function. A well designed waveform with 4GHz bandwidth will give a peak with 3dB width corresponding to a range resolution of about 5cm and sidelobe levels at least 40dB down from the peak.

The apparatus described above has a single range gate whose position is determined by the delay in the path of the transmission waveform sample and whose width is determined by the range resolution. However this range gate may be swept through the range swath of interest by varying the delay. Alternatively a multi range gate system can be constructed by dividing the sample into a number of different delays and feeding each to an identical receiver channel.

Although in this embodiment only two mixers for multiplying bandwidth are shown, more than two mixers may be used in series to increase the bandwidth progressively by a factor of two at each mixer stage. In this way the system may achieve ultra high range resolution. For example, a two stage system will produce a 4GHz bandwidth and about 5cm resolution and a three stage system will produce a 8GHz bandwidth and about 2.5cm resolution. With a greater number of mixers, sub-1cm resolution may be achievable. Further processing of the signal may be required after additional mixer stages, for example, filtering and amplification.

Although in the example described each mixer doubles the frequency and bandwidth of the signals, it may be desirable to increase the bandwidth whilst maintaining the frequency within a specific or desirable range. In this event further mixers may be used to mix the signal with a fixed local oscillator signal to step down the signals. However the bandwidth would be maintained.

Rather than using different time delays in order to step the range gate, the system may comprise multiple receiver channels having different delays.

The invention may also be used to enhance and improve imaging of targets by radar and may be used for high resolution tracking of moving objects.

20

15

5

10

15

20

CLAIMS

- 1. A radar apparatus comprising waveform generator means for generating a signal at an intermediate frequency, a coupler for dividing the signal into two divided signals, a mixer for mixing the two divided signals together to generate an output signal and transmission means for transmitting the output signal as a radar signal in which the output signal has a wider bandwidth than the signal at the intermediate frequency.
- 2. Apparatus according to claim 1 in which the frequency of the output signal is twice the intermediate frequency of the signal.
 - 3. Apparatus according to claim 1 or claim 2 in which a path between a first output of the coupler and a first input of the mixer and a path between a second output of the coupler and a second input of the mixer are of equal length, or of equal phase length.
 - Apparatus according to any preceding claim in which the output signal is sent to one or more additional coupler and mixer combinations to increase the bandwidth further.
 - 5. Apparatus according to any preceding claim in which there are a plurality of mixers.
 - 6. Apparatus according to claim 5 in which there are two mixers.
 - Apparatus according to any preceding claim in which the intermediate frequency is at
 1 GHz or higher.

- 8. Apparatus substantially as described herein with reference to the Figure.
- A radar detection system comprising a radar apparatus according to any preceding claim and means for receiving and processing radar signals.

15

- 10. A system according to claim 9 in which the means for receiving radar signals is a correlation receiver, in which a received signal reflected from a target is correlated with a reference signal.
- 10 11. A system according to claim 10 in which the reference signal is a delayed sample of the output signal which was transmitted.
 - 12. A system according the claims 10 or 11 in which correlation is performed by mixing the received signal and the reference signal in a wideband mixer and integrating the mixer output for a time corresponding to the length of the transmitted waveform.
 - 13. A system substantially as described herein with reference to the Figure.
 - 14. A method of generating a radar signal comprising the steps of:
- 20 generating a signal having an intermediate frequency;

dividing the signal into two divided signals;

mixing the divided signals to generate an output signal having a wider bandwidth than

the intermediate frequency signal; and

transmitting the output signal as a radar signal.

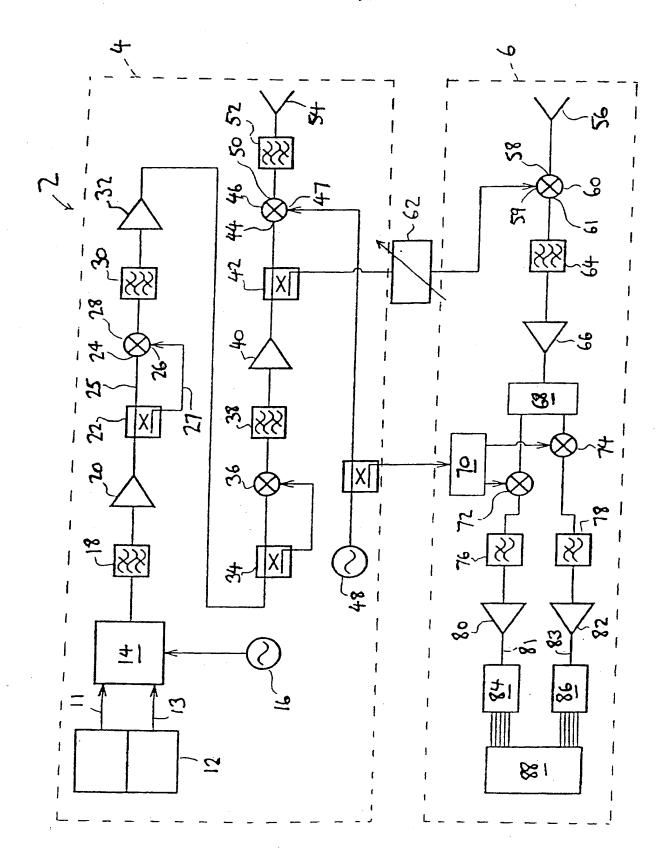
15. A method substantially as described herein with reference to the Figure.

5

10

15

- 20



INTERNATIONAL SEARCH REPORT

In Atlanta Application No PCT/GB 97/02513

A. CLASSIFICATION OF SUBJECT MATTER IPC 6 G01S7/282						
According to International Patent Classification (IPC) or to both national classification and IPC						
	SEARCHED	too cymbolo)				
	Minimum documentation searched (classification system followed by classification symbols) IPC 6 G01S H03C H03B					
Documenta	ation searched other than minimum documentation to the extent that	such documents are included in the fields sea	rched			
		·				
Electronic	data base consulted during the international search (name of data b	ase and, where practical, search terms used)				
C. DOCUM	ENTS CONSIDERED TO BE RELEVANT					
Category *	Citation of document, with indication, where appropriate, of the re	levant passages	Relevant to claim No.			
Υ	US 5 151 661 A (CALDWELL STEPHEN 29 September 1992 see the whole document	P ET AL)	1,2,4,7, 14			
Υ	US 5 077 546 A (CARFI RALPH W E December 1991 see column 4 - column 5; figure		1,2,4,7,			
Α	DRISCOLL M M ET AL: "SPECTRAL PERFORMANCE OF FREQUENCY MULTIPLIERS AND DIVIDERS" PROCEEDINGS OF THE FREQUENCY CONTROL SYMPOSIUM, HERSHEY, MAY 27 - 29, 1992, no, 27 May 1992, INSTITUTE OF ELECTRICAL AND ELECTRONICS ENGINEERS, pages 193-200, XP000341357 see page 195, left-hand column, last line - right-hand column, line 1					
		-/				
X Furti	her documents are listed in the continuation of box C.	Patent family members are listed in	annex.			
"A" docume consid	ont which may throw doubts on priority claim(s) or is cited to establish the publication date of another nor or other special reason (as specified) and the priority of the referring to an oral disclosure, use, exhibition or	"T" later document published after the international filing date or priority date and not in conflict with the application but cited to understand the principle or theory underlying the invention "X" document of particular relevance; the claimed invention cannot be considered novel or cannot be considered to involve an inventive step when the document is taken alone "Y" document of particular relevance; the claimed invention cannot be considered to involve an inventive step when the document is combined with one or more other such documents, such combination being obvious to a person skilled in the art. "&" document member of the same patent family				
	actual completion of theiriternational search December 1997	Date of mailing of the international search report 09/12/1997				
		Authorized officer				
Name and mailing address of the ISA European Patent Office, P.B. 5818 Patentlaan 2 NL - 2280 HV Rijswijk Tel. (+31-70) 340-2040, Tx. 31 651 epo nl, Fax: (-31-70) 340-3016		Zaccà, F	·			

Form PCT/ISA/210 (second sheet) (July 1992)

1

INTERNATIONAL SEARCH REPORT

in national Application No PCT/GB 97/02513

		PCT/GB 97/02513			
	inuation) DOCUMENTS CONSIDERED TO BE RELEVANT y * Citation of document, with indication, where appropriate, of the relevant passages Relevant to claim No.				
Category °	Chancer of document, with indication, where appropriate, or the relevant passages	Helevani to daim No.			
	PATENT ABSTRACTS OF JAPAN vol. 096, no. 004, 30 April 1996 & JP 07 336145 A (FUKUSHIMA NIPPON DENKI KK), 22 December 1995, see abstract	1			
		•			

ī

Form PCT/ISA/210 (continuation of second sheet) (July 1992)

INTERNATIONAL SEARCH REPORT

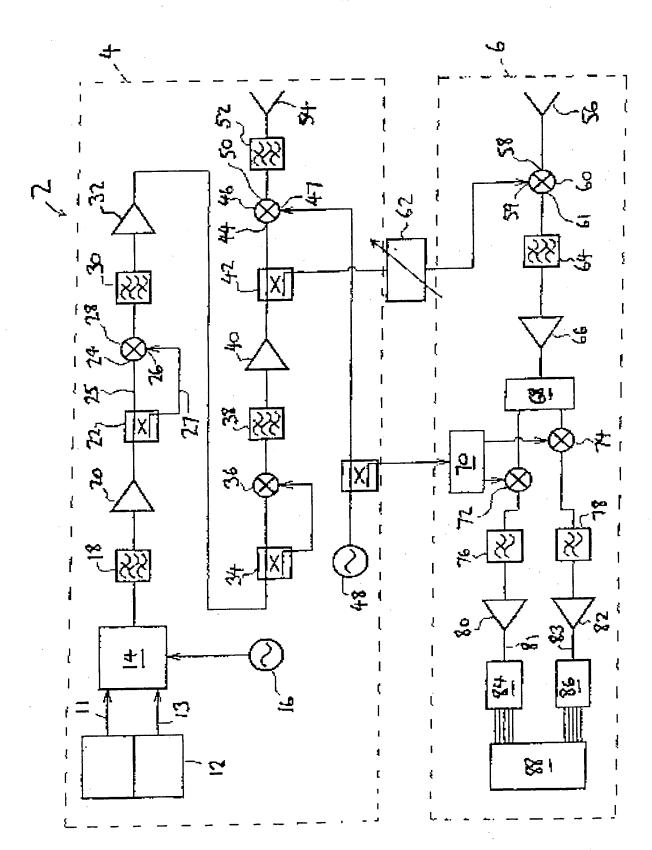
information on patent family members

International Application No PCT/GB 97/02513

Patent document cited in search report	Publication date	Patent family member(s)	Publication date
US 5151661 A	29-09-92	NONE	
US 5077546 A	31-12-91	CA 2048762 A DE 69104167 D DE 69104167 T EP 0485050 A	08-05-92 27-10-94 11-05-95 13-05-92

Form PCT/ISA/210 (patent family annex) (July 1992)

BNSDOCID: <WO_____9813704A1_I_>



BEST AVAILABLE COPY

THIS PAGE BLANK (USPTO)